

# Model-based Predictive Control (MPC)

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SCM-MPC Workshop, München



Technische Universiteit **Eindhoven** University of Technology

Where innovation starts

## My background

#### (Nonlinear) Control Theory





#### What I like

Apply ideas/concepts from control theory in other fields



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## Four major problems

- 1. Generate feasible reference trajectory
- 2. Design (static) state feedback controller
- 3. Design observer
- 4. Design (dynamic) output feedback controller



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## Example: Linear dynamics, discrete time

$$\begin{aligned} x(k+1) &= Ax(k) + Bu(k) \qquad x(0) = x_0 \\ y(k) &= Cx(k) \end{aligned}$$

Problem 1: Generate feasible reference trajectory

Determine  $(u_r(k), x_r(k), y_r(k))$  satisfying

$$x_r(k+1) = Ax_r(k) + Bu_r(k)$$
$$y_r(k) = Cx_r(k)$$

Typical approach: solve optimization problem



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## Design (static) state feedback controller

#### Given dynamics and reference

 $x(k+1) = Ax(k) + Bu(k) \qquad x_r(k+1) = Ax_r(k) + Bu_r(k)$ 

Define tracking error and change of input:

 $\tilde{x}(k) = x(k) - x_r(k)$   $\tilde{u}(k) = u(k) - u_r(k)$ 

Resulting in error dynamics:

$$\tilde{x}(k+1) = A\tilde{x}(k) + B\tilde{u}(k)$$
  
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## Design (static) state feedback controller

Problem: Determine  $\tilde{u}([\tilde{x}(k)])$  such that  $\lim_{k\to\infty} \tilde{x}(k) = 0$ , where

$$\tilde{x}(k+1) = A\tilde{x}(k) + B\tilde{u}(k)$$
  $\tilde{x}(0) = \tilde{x}_0$ 

#### Solution

Use  $\tilde{u}(k) = -L\tilde{x}(k)$ . Resulting closed loop dynamics:

 $\tilde{x}(k+1) = (A - BL)\tilde{x}(k)$   $x(0) = \tilde{x}_0$ 

# Lemma: If rank $\begin{bmatrix} B & AB & A^2B & \dots & A^{n-1}B \end{bmatrix} = n$ then eigenvalues of A - BL can be placed arbitrarily.



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Given dynamics and reference

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we use the following controller:

$$u(k) = u_r(k) - L[x(k) - x_r(k)]$$

which guarantees that  $\lim_{k\to\infty} x(k) - x_r(k) = 0$ , provided *L* is properly chosen.



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## Design observer

#### Given dynamics

$$\begin{aligned} x(k+1) &= Ax(k) + Bu(k) \qquad x(0) = x_0 \\ y(k) &= Cx(k) \end{aligned}$$

#### Is it possible to reconstruct x(k) from u(k) and y(k)?

#### Solution

Let  $\hat{x}(k)$  denote our estimate for x(k):

$$\hat{x}(k+1) = A\hat{x}(k) + Bu(k)$$
$$\hat{y}(k) = C\hat{x}(k)$$

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#### Solution

Let  $\hat{x}(k)$  denote our estimate for x(k):

$$\hat{x}(k+1) = A\hat{x}(k) + Bu(k) + K[y(k) - \hat{y}(k)]$$
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## Design observer

Given dynamics and observer

$$x(k+1) = Ax(k) + Bu(k)$$
$$y(k) = Cx(k)$$

 $x(0) = x_0$ 

$$\hat{x}(k+1) = A\hat{x}(k) + Bu(k) + K[y(k) - \hat{y}(k)]$$
$$\hat{y}(k) = C\hat{x}(k)$$

For observer error  $\bar{x}(k) = x(k) - \hat{x}(k)$ , we obtain

$$\bar{x}(k+1) = A\bar{x}(k) - K\bar{y}(k) = [A - KC]\bar{x}(k)$$
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$$x(k+1) = Ax(k) + Bu(k)$$
$$y(k) = Cx(k)$$

 $x(0)=x_0$ 

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## Design (dynamic) output feedback controller

Given dynamics and reference

 $x(k+1) = Ax(k) + Bu(k) \qquad x_r(k+1) = Ax_r(k) + Bu_r(k)$ 

we use the following controller:

$$u(k) = u_r(k) - L[\hat{x}(k) - x_r(k)]$$
$$\hat{x}(k+1) = A\hat{x}(k) + Bu(k) + K[y(k) - \hat{y}(k)]$$
$$\hat{y}(k) = C\hat{x}(k)$$

which guarantees that  $\lim_{k\to\infty} x(k) - x_r(k) = 0$ , provided K and L are properly chosen.



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# Introduction to Control Theory: Summary

#### Four major problems

- 1. Generate feasible reference trajectory
- 2. Design (static) state feedback controller
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#### System

$$X(k+1) = A(k)X(k) + B(k)U(k) + V(k)$$
$$Y(k) = C(k)X(k) + W(k)$$

## with V(k), W(k) Gaussian white noise (cov. matrices $\Sigma_v(k)$ , $\Sigma_w(k)$ ).

#### Objective

Minimize

$$J = E\left(X(N)^{T}Q(N)X(N) + \sum_{k=0}^{N-1} X(k)^{T}Q(k)X(k) + U(k)^{T}R(k)U(k)\right)$$

where 
$$Q(k) \ge 0$$
,  $R(k) > 0$ .

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where  $Q(k) \ge 0$ , R(k) > 0.





## Solution

Controller

$$U(k) = -L(k)\hat{X}(k)$$
$$\hat{X}(k+1) = A(k)\hat{X}(k) + B(k)U(k) + K(k)[Y(k) - C(k)\hat{X}(k)]$$



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## Solution

Controller

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$$K(k) = A(k)P(k)C(k)^{T}[C(k)P(k)C(k)^{T} + \Sigma_{w}(k)]^{-1}$$
  

$$P(k+1) = A(k)\Gamma[C(k), P(k), \Sigma_{w}(k)]A(k)^{T} + \Sigma_{v}(k)$$
  

$$L(k) = [B(k)^{T}S(k+1)B(k) + R(k)]^{-1}B(k)^{T}S(k+1)A(k)$$
  

$$S(k) = A(k)^{T}\Gamma[B(k)^{T}, S(k+1), R(k)]A(k) + Q(k)$$
  

$$(N) = Q(N), P(0) = E(X_{0}X_{0}^{T}), \hat{x}(0) = E(X_{0}).$$
 Furthermore,  

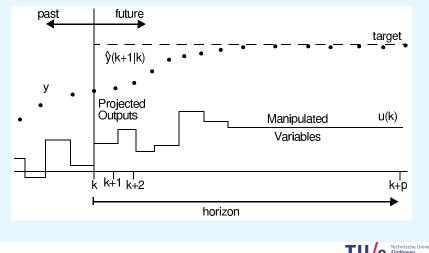
$$\Gamma[F, G, H] = G - GF^{T}(FGF^{T} + H)^{-1}FG$$



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# Model based predictive control

## **Receding horizon**



#### First era: industrial success stories

- 1950s Various oil and petrochemical industries: optimal process settings computed every 15-20 minutes, implemented by manual operators.
- 60s and 70s Feedback controller from repeatedly using recomputed open loop controllers (Lee and Markus). Repeatedly solving Problem 1.
  - Deterministic (without any disturbance model)
  - Lack of stability guarantees
  - Lack of systematic tuning guidelines



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## Second era: founding of MPC theory

Consider system x(k + 1) = Ax(k) + Bu(k). At each time: measure (or estimate) state  $x_0$  and solve

$$\min_{\substack{u(0),\dots,u(N-1)}} \hat{x}(N)^T Q_N x(N) + \sum_{k=0}^{N-1} \left[ x(k)^T Q x(k) + u(k)^T R u(k) \right]$$
  
s.t.  $x(k+1) = A x(k) + B u(k)$   $x(0) = x_0$   
 $u(k) \in \mathbb{U}$   
 $x(k) \in \mathbb{X}$   $x(N) \in \mathbb{X}_N$ 

where  $\mathbb{U}$ ,  $\mathbb{X}$ ,  $\mathbb{X}(p)$  convex compact sets containing 0.

Result: feedback  $u(x_0)$  (online calculation).



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# Model based predictive control

#### Example

## **Consider dynamics**

x(k+1) = 4x(k) + u(k)  $x(k|k) = x_0$ 

Horizon of 1:  $\min_{u(k|k)} x(k+1|k)^2 + u(k|k)^2$ 

 $\min_{u(k|k)} [4x_0 + u(k|k)]^2 + u(k|k)^2 = 16x_0^2 + 8x_0u(k|k) + 2u(k|k)^2$ 

Optimal solution:  $u(k|k) = \frac{-8x_0}{2\cdot 2} = -2x_0$ 

Closed-loop system: x(k + 1) = 4x(k) - 2x(k) = 2x(k) Unstable!



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#### Observation 1: Infinite horizon results in stabilizing controller

Observation 2: After finite amount of time: solution remains unconstrained

Idea: Properly select terminal costs and horizon

#### Main steps

- 1. Solve infinite horizon LQR problem: u = Kx,  $V = x^T Px$
- 2. Determine maximal output admissible set:  $X_N$
- 3. Determine N s.t.  $x(N) \in \mathbb{X}_N$



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# Second era: founding of MPC theory

- Robust MPC (next slide)
- Nonlinear MPC

## Third era: Diversification through fast MPC

- MPC for hybrid systems and systems with logical constraints
- Explicit MPC (mpLP,mpQP)
- Fast optimization
- Application (mechanical and electronic systems)



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## **Robust MPC**

### Dynamics

 $x(k+1) = Ax(k) + Bu(k) + Ed(k) \qquad d(k) \in \mathbb{D} = \{d : Ld \leq l\}$ 

$$J^{*(k)}(x^{(k)}) = \min_{u^{(k)}} J^{(k)}(x^{(k)}, u^{(k)})$$
  
s.t.  $\begin{cases} Fx^{(k)} + Gu^{(k)} \le g \\ Ax^{(k)} + Bu^{(k)} \in \mathbb{X}^{(k)} \end{cases}$   $\forall d^{(k)} \in \mathbb{D}$   
 $J^{(k)}(x^{(k)}, u^{(k)}) = \max_{d^{(k)} \in \mathbb{D}} ||Qx^{(k)}||_1 + ||Ru^{(k)}||_1 + J^{*(k+1)}(Ax^{(k)} + Bu^{(k)} + Ed^{(k)})$   
 $\mathbb{X}^{(k)} = \{x \in \mathbb{R}^n : \forall d \in \mathbb{D} \; \exists u \in \mathbb{R}^{n_u} \text{ with}$   
 $Fx + Gu \le g \text{ and } Ax + Bu + Ev \in \mathbb{X}^{(k+1)} \}.$   
where  $J^{*K}(x^{(K)}) = 0$  and  $\mathbb{X}^{(K)} = \{x \in \mathbb{R}^n : Fx \le g\}$ 

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 $\mathbb{X}^{(k)} = \{x \in \mathbb{R}^n : \forall d \in \mathbb{D} \; \exists u \in \mathbb{R}^{n_u} \text{ with}$   
 $Fx + Gu \le g \text{ and } Ax + Bu + Ev \in \mathbb{X}^{(k+1)} \}.$ 

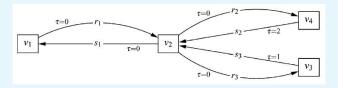
#### **Robust MPC**

#### Dynamics

 $x(k+1) = Ax(k) + Bu(k) + Ed(k) \qquad d(k) \in \mathbb{D} = \{d : Ld \leq l\}$ 

$$J^{*(k)}(x^{(k)}) = \min_{u^{(k)}} J^{(k)}(x^{(k)}, u^{(k)})$$
  
s.t.  $\left\{ \begin{matrix} Fx^{(k)} + Gu^{(k)} \leq g \\ Ax^{(k)} + Bu^{(k)} \in \mathbb{X}^{(k)} \end{matrix} \right\} \forall d^{(k)} \in \mathbb{D}$   
 $J^{(k)}(x^{(k)}, u^{(k)}) = \max_{d^{(k)} \in \mathbb{D}} ||Qx^{(k)}||_1 + ||Ru^{(k)}||_1 + J^{*(k+1)}(Ax^{(k)} + Bu^{(k)} + Ed^{(k)})$   
 $\mathbb{X}^{(k)} = \{x \in \mathbb{R}^n : \forall d \in \mathbb{D} \; \exists u \in \mathbb{R}^{n_u} \text{ with}$   
 $Fx + Gu \leq g \text{ and } Ax + Bu + Ev \in \mathbb{X}^{(k+1)} \}.$   
where  $J^{*K}(x^{(K)}) = 0$  and  $\mathbb{X}^{(K)} = \{x \in \mathbb{R}^n : Fx \leq g\}$ 

## Robust MPC: Example



Retailer  $v_2$ : uncertain demand  $d(t) \in [0, 8]$ 

- Order  $u_1(t) \in [0, 6]$  from supplier  $v_3$ : cost 4, delay 1
- Order  $u_2(t) \in [0, 6]$  from supplier  $v_4$ : cost 1, delay 2



/department of mechanical engineering

#### Robust MPC: Example

$$x(t+1) = \begin{bmatrix} 1 & 1 & 0 & 1 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} x(t) + \begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 1 & 0 \\ 0 & 1 \end{bmatrix} u(t) + \begin{bmatrix} -1 \\ 0 \\ 0 \\ 0 \end{bmatrix} d(t).$$

Resulting (dual base stock) policy:

$$u_1^*(x) = \min\{\max\{20 - x_1 - x_2 - x_3 - x_4, 0\}, 4\}, u_2^*(x) = \max\{16 - x_1 - x_2 - x_3 - x_4, 0\}.$$

